

## 3.2 Orthogonal Coordinate Systems

#### **Cartesian Coordinates**

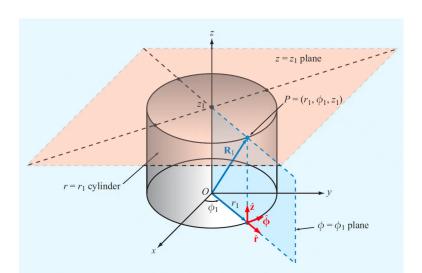
 $d\mathbf{l} = \hat{x} \ dx + \hat{y} \ dy + \hat{z} \ dz$ 

#### Cylindrical Coordinates

- Measured in  $r, \Phi, z$ .
  - $\circ~\Phi$  is the azimuth angle measured counterclockwise from the positive x axis in the x-y plane

 $dV = r dr d\Phi dz$ 

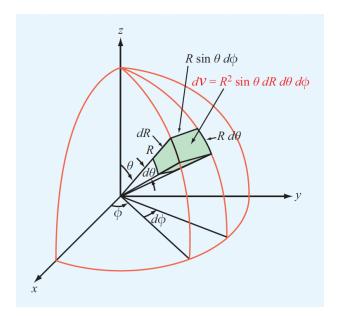
• Differential volume element is given in:



#### Cylindrical Coordinates

- Measured in  $R, heta, \Phi$ 
  - $\circ$  The zenith angle heta is measured from positive z-axis downwards
- Differential volume element is given by:

$$dV=R^2~\sin heta~dR~d heta~d\Phi$$



## 3.4 Gradient of a Scalar Field

#### Gradient

• In Cartesian coordinates

$$abla = \hat{x}rac{\partial}{\partial x} + \hat{y}rac{\partial}{\partial y} + \hat{z}rac{\partial}{\partial z}$$

• In cylindrical coordinates

$$abla = \hat{r} rac{\partial}{\partial r} + \hat{\Phi} rac{1}{r} rac{\partial}{\partial \Phi} + \hat{z} rac{\partial}{\partial z}$$

• In spherical coordinates

$$abla = \hat{R}rac{\partial}{\partial R} + \hat{ heta}rac{1}{R}rac{\partial}{\partial heta} + \hat{\Phi}rac{1}{R\sin heta}rac{\partial}{\partial \Phi}$$

#### Properties of the Gradient Operator

$$abla (U+V) = 
abla U + 
abla V$$
 $abla (UV) = U
abla V + V
abla U$ 
 $abla V^n = nV^{n-1}
abla V$ 

## 3.7 The Laplacian Operator

- For a vector  ${\bf E}$  specified in Cartesian coordinates, the Laplacian of  ${\bf E}$  is given by:

$$abla^2 {f E} = igg( {\partial^2 \over \partial x^2} + {\partial^2 \over \partial y^2} + {\partial^2 \over \partial z^2} igg) {f E}$$

• Through direct substitution, it can also be shown that (relevance unknown)

$$abla^2 {f E} = 
abla (
abla \cdot {f E}) - 
abla imes (
abla imes {f E})$$



## 4.1 Maxwell's Equations

• Modern electromagnetic theory is based on four fundamental relations known as Maxwell's equations

$$egin{aligned} 
abla \cdot \mathbf{D} &= 
ho_v \ 
abla imes \mathbf{E} &= -rac{\partial \mathbf{B}}{\partial t} \ 
abla imes \mathbf{B} &= 0 \ 
abla imes \mathbf{H} &= \mathbf{J} + rac{\partial \mathbf{D}}{\partial t} \end{aligned}$$

- ${\bf E}$  and  ${\bf D}$  are the electric field intensity and flux density
  - $\circ$  Correlated by  $\mathbf{D} = arepsilon \mathbf{E}$  where arepsilon is the **electrical permittivity**
- ${\bf H}$  and  ${\bf B}$  are the magnetic field intensity and flux density
  - Correlated by  $\mathbf{B} = \mu \mathbf{H}$  where  $\mu$  is the magnetic permeability
- James Clerk Maxwell published these equations in 1873 and established the first unified theory of electricity and magnetism

Under **static** conditions, all functions of time go to zero

#### Electrostatics

$$abla \cdot {f D} = 
ho_{i} 
onumber 
on$$

#### Magnetostatics

$$abla \cdot \mathbf{B} = 0$$
 $abla \times \mathbf{H} = \mathbf{J}$ 

• We can study electricity and magnetism as separate phenomena so long as distributions of charge and current flow stay constant

#### **4.2 Charge and Current Distributions**

#### **Charge Densities**

- Volume charge density  $ho_v$ 

$$ho_v = \lim_{\Delta v o 0} rac{\Delta q}{\Delta V} = rac{dq}{dV} ~~(C/m^3)$$

- The total charge contained in a volume  $\boldsymbol{V}$  is

$$Q=\int_v 
ho_v \, dV ~~(C)$$

• The surface charge density  $ho_s$  is given by:

$$ho_s = \lim_{\Delta s o 0} rac{\Delta q}{\Delta s} = rac{dq}{ds} ~~(C/m^2)$$

• Line charge density  $\rho_l$  is given by:

$$ho_l = \lim_{\Delta l o 0} rac{\Delta q}{\Delta l} = rac{dq}{dl} ~~(C/m)$$

#### **Current Density**

- Let  ${\bf u}$  be the velocity at which charges move in a tube. Then, the current density is given by:

$$\mathbf{J}=
ho_v\mathbf{u}~~(A/m^2)$$

• Then, the total current flowing through a surface is

$$I = \int_S {f J} \cdot d{f s} ~~(A)$$

When a current is generated by actual movement of charged matter, it is called convection current, and  ${\bf J}$  is called a convectional current density.

Otherwise, if the current is generated by movement of charged particles relative to the host material, we call it **conduction current**.

#### 4.3 Coulomb's Law

- Coulomb's Law was first introduced for electrical charges in air, and was later generalized to other media
- Coulomb's Law implies that:
  - $\circ$  An isolated charge q induces an electric field  ${f E}$  at every point in space, where  ${f E}$  is given by

$${f E}=rac{q}{4\piarepsilon R^2}\hat{R}~~(V/m)$$

 $\circ$  In the presence of an electric field  ${\bf E}$  at any given point in space, the force acting on a small, positive test charge is

$$\mathbf{F} = q\mathbf{E}$$

#### Electric Field Due to Multiple Point Charges

• The electric field at any given point is the vector sum of the field caused by all point charges

#### Electric Field Due to Charge Distribution

• Volume distrubtion

$${f E}={1\over 4\piarepsilon}\int_V {
ho_v\over R^2} \, dV$$

• Surface distribution

$${f E}=rac{1}{4\piarepsilon}\int_{S}rac{
ho_{s}}{R^{2}}\,dS\,,$$

• Line distribution

$${f E}={1\over 4\piarepsilon}\int_l {
ho_l\over R^2}~dl$$

• For an infinite sheet of charge we have

$$\mathbf{E}=rac{
ho_{s}}{2arepsilon_{0}}$$

#### 4.4 Gauss' Law

• We begin by restating the differential form of Gauss' Law:  $abla \cdot {f D} = 
ho_v$ 

$$\int_v 
abla \cdot \mathbf{D} \; dV = \oint_S \mathbf{D} \cdot d\mathbf{s} = Q$$

- Maxwell's equations incorporate Gauss' law in themselves
  - For a simple case such as an isolated point charge, we can use Coulomb's law
  - For more complex systems, we can still use Coulomb's law, but Gauss' law is much easier to apply
    - A shortcoming is that it can only be applied to symmetrical charge distributions

#### **4.5 Electric Scalar Potential**

- Operation of an electric circuit usually described in terms of currents flowing through branches and voltage at nodes.
- Voltage difference V b/w two nodes represents the amount of work or **potential** energy required to move a unit charge from one terminal to the other.
- Subject of this section is relationship between  $ec{E}$  and V.

#### Electric Potential as a Function of Electric Field

• When a charged particle is in an electric field there is

$$ec{F}_{ext} = -qec{E}$$

- Work done in moving any object a vector differential distance  $d\vec{l}$  while exerting a force  $\vec{F}_{ext}$  is:

$$dW = ec{F}_{ext} \cdot dec{l} = -qec{E} \cdot dec{l}$$

- Differential electric potential energy dW per unit charge is called the differential electric potential dV. That is,

$$dV = rac{dW}{q} = -ec{E} \cdot dec{l}$$

• Units are (J/C) or (V)  $% \left( V^{\prime}\right) =0$ 

$$V_{21} = V_2 - V_1 = -\int_{P_1}^{P_2}ec{E}\cdot dec{l}$$

The voltage difference between two nodes in an electric circuit has the same value regardless of which path in the circuit we follow in between the nodes. Moreover, Kirchoff's voltage law states the net voltage drop around a closed loop is zero.

- The line integral of the electrostatic field  $ec{E}$  around any closed contour C is 0.
- Conservative property of the electrostatic field can be deduced from Maxwell's second equation. If  $\partial/\partial t=0$ , then

$$abla imes ec E = 0$$

- If we integrate this over an open surface  ${\cal S}$  and apply Stokes' Theorem to convert the surface integral into a line integral, we obtain

$$\int_S (
abla imes ec E) \cdot dec s = \oint_C ec E \cdot dec l = 0$$

#### Electric Potential Due to Point Charges

Electric field due to a point charge q is given by:

$$ec{E}=\hat{R}rac{q}{4\piarepsilon R^2}$$

#### Electric Potential Due to Continuous Distributions

Volume distribution

$$V = rac{1}{4\piarepsilon}\int_{v'}rac{
ho_v}{R'}\;dv'$$

Charge distribution

$$V = rac{1}{4\piarepsilon} \int_{S'} rac{
ho_s}{R'} \ ds'$$

Line distribution

$$V = rac{1}{4\piarepsilon}\int_{l'}rac{
ho_l}{R'}\; dl'$$

#### Electric Field as a Function of Electric Potential

 $\vec{E}=-\nabla V$ 

This differential relationship between V and  $ec{E}$  allows us to determine  $ec{E}$  for any charge distribution by first calculating V and then taking the negative gradient of V.

- An electric dipole consists of two point charges, equal magnitude but opposite polarity separated by a distance d.
- The dipole moment is given by  $\vec{p} = q\vec{d}$ . Then, we have:

$$V = rac{ec{p} \cdot \hat{R}}{4\piarepsilon_0 R^2}$$

#### **Poisson's Equation**

$$abla \cdot ec{E} = rac{
ho_v}{arepsilon}$$

$$abla^2 V = -rac{
ho_v}{\epsilon} ext{ (Poisson's equation)}$$

#### 4.6 Conductors

- A material medium has electromagnetic constitutive parameters:
  - $\circ$  Electrical permittivity  $\varepsilon$
  - $\circ$  Magnetic permeability  $\mu$
  - $\circ$  Conductivity  $\sigma$
- Homogeneous means the constitutive parameters do not vary by position
- Isotropic means the constitutive parameters do not vary from point to point
- Conduction current density is given by:

$$\vec{J} = \sigma \vec{E}$$

- A perfect dielectric has  $\sigma=0$  and a perfect conductor has  $\sigma=\infty$ 

#### Drift Velocity

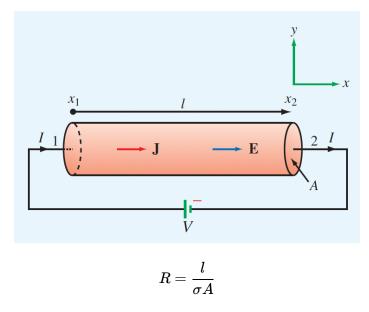
$$ec{u}_e = -\mu_eec{E}~({
m m/s})$$

- $\mu_e$  is a property called the electron mobility. Similarly, we have hole drift velocity and hole mobility
- The total conduction current density is given by:

$$egin{aligned} ec{J} &= ec{J}_e + ec{J}_h = 
ho_{ve}ec{u}_e + 
ho_{vh}ec{u}_h ~( ext{A/m^2}) \ ec{J} &= (-
ho_{ve}\mu_e + 
ho_{vh}\mu_h)ec{E} = \sigmaec{E} \end{aligned}$$

- For a perfect dielectric we would have  $ec{J}=0$  and  $ec{E}=0$ 

#### Resistance



- Reciprocal of R is called G which is conductance, with a unit of  $\Omega^{-1}.$  For a linear resistor:

$$G = \frac{1}{R} = \frac{\sigma A}{l}$$

• If  $\vec{J}$  is in the  $\hat{r}$  direction, the inner conductor must be at a higher potential than the outer conductor. The voltage difference is given by:

$$V_{ab} = -\int_{b}^{a}ec{E}\cdot dec{l} = -\int_{b}^{a}rac{I}{2\pi\sigma l}rac{\hat{r}\cdot\hat{r}\;dr}{r} = rac{I}{2\pi\sigma l}\ln\left(rac{b}{a}
ight)$$

#### Joule's Law

• The work expended by the electric field in moving  $q_e$  a differential distance  $\Delta l_e$  and moving a  $q_h$  a distance  $\Delta l_h$  is:

$$\Delta W = ec{F}_e \cdot \Delta ec{l}_e + ec{F}_h \cdot \Delta ec{l}_h$$

• Power P is measured in units of watts (W) and is defined as the time rate of change in energy. For a volume V, the total dissipated power is:

$$P = \int_v ec{E} \cdot ec{J} \; dV \; ext{(Joule's Law)}$$

#### **4.7 Dielectrics**

• In a conductor electrons are loosely bound to their atom, whereas in dielectrics the atoms are tightly bound

#### **Polarization Field**

In a free space with  $\vec{D} = \varepsilon_0 \vec{E}$ , the presence of microscopic dipoles in a dielectric material alters that relationship to:

$$\vec{D} = \varepsilon_0 \vec{E} + \vec{P}$$

where  $\vec{P}$  is the electric polarization field.  $\vec{P}$  is directly proportional to  $\vec{E}$  and is expressed as

$$ec{P}=arepsilon_0\chi_eec{E}$$

where  $\chi_e$  is the electric susceptibility of the material.

Permittivity of a material  $\varepsilon$  is given by:

$$arepsilon = arepsilon_e (1+\chi_e)$$

#### **Dielectric Breakdown**

The preceding model assumes that the magnitude  $\vec{E}$  will not exceed a certain critical value, the dielectric strength  $\vec{E}_{ds}$ . Beyond this, electrons will detach and accelerate though the material as a conduction current. This is known as dielectric behaviour.

$$V_{br} = E_{ds}d$$

#### **4.8 Electric Boundary Conditions**

A vector field is spatially continuous if it does not exhibit abrupt changes in either magnitude of direction when expressed as a function of position.

At the boundary of two distinct media, one notices that:

$$(ec{E}_1-ec{E}_2)\cdot \hat{l}_1=0$$

I other words, the tangential component of the electric field is continuous across the boundary between any two media.

$$\vec{E}_{1t} = \vec{E}_{2t}$$

#### **Dielectric-Conductor Boundary**

If medium 1 is a dielectric and medium 2 is a perfect conductor, then because the electric fields and fluxes vanish in a conductor, it follows that  $\vec{E}_2 = \vec{D}_2 = 0$ . This implies that the tangential and normal components to the interface are both zero.

#### **Conductor-Conductor Boundary**

If medium 1 has permittivity  $\varepsilon_1$  and conductivity  $\sigma_1$ , and medium 2 has permittivity  $\varepsilon_2$  and conductovity  $\sigma_2$ , then the interface between them holds a surface charge density  $\rho_s$ .

$$ec{E}_{1t} = ec{E}_{2t} ext{ and } arepsilon_1 E_{1n} - arepsilon_2 E_{2n} = 
ho_s$$

The normal component of  $\vec{J}$  has to be continuous across the boundary between two different media under electrostatic conditions.

$$J_{1n}igg(rac{arepsilon_1}{\sigma_1}-rac{arepsilon_2}{\sigma_2}igg)=
ho_s$$

#### 4.9 Capacitance

- When separated by a dielectric, any two conducting bodies form a capacitor.
- If a DC voltage is connected across the surfaces, the positive and negative source terminals accumulate charges of +Q and -Q respectively.

When a conductor has excess charge, it distributes the charge on its surface to maintain a zero electric field everywhere within the conductor.

$$C=rac{Q}{V}\left(\mathrm{C/V~or~F}
ight)$$

• The tangential component of  $\vec{E}$  always vanishes at a conductor's surface, so  $\vec{E}$  is always perpendicular. The normal component is then given by:

$$E_n = \hat{n} \cdot ec{E} = rac{
ho_s}{arepsilon}$$

• The charge Q is equal to the integral of  $ho_s$  over the surface S.

$$egin{aligned} Q &= \int_{S} arepsilon ec{E} \cdot dec{S} \ C &= rac{\int_{S} arepsilon ec{E} \cdot dec{s} \ - \int_{l} ec{E} \cdot dec{l} \end{aligned}$$

The value of C obtained for any specific capacitor configuration is always independent of the magnitude of  $\vec{E}$ .

If the material between the conductors is not a perfect dielectric but has a small conductivity  $\sigma$ , then the general expression for the R resistance is:

$$R = \frac{-\int_{l} \vec{E} \cdot d\vec{l}}{\int_{S} \sigma \vec{E} \cdot d\vec{s}}$$

For a uniform conductivity and permittivity, we then obtain

$$RC = rac{arepsilon}{\sigma}$$

The voltage difference between the plates is:

$$V=-\int_0^dec{E}\cdot dec{l}=i\int_0^d(-\hat{z}E)\cdot\hat{z}\ dz=Ed$$

And the capacitance would then be:

$$C = rac{Q}{V} = rac{Q}{Ed} = rac{arepsilon A}{d}$$

#### **4.10 Electrostatic Potential Energy**

- The energy spent in charging a capacitor using a power supply us stored in the dielectric medium in the form of electrostatic potential energy.
- The volage v across a capacitor is related to the charge stored q by

$$v = rac{q}{C}$$

• The amount of work W required to charge the capacitor can be given by

$$W_e=rac{1}{2}CV^2$$

where V is the final voltage.

- The electrostatic energy density  $w_e$  is defined as the electrostatic potential energy  $W_e$  per unit volume:

$$w_e = rac{W_e}{
u} = rac{1}{2}arepsilon E^2$$

• The opposing charged plates are also attracted to each other by an electrical force

$$ec{F}_e = -\hat{z}F_e$$

where  $F_e$  is given by:

$$F_e=rac{1}{2}arepsilonrac{AV^2}{d^2}$$

• We generalize this result for any  $dec{l}$  along any direction as:

$$ec{F}_e = -
abla W_e$$



## Magnetostatics

#### **5.1 Magnetic Forces and Torques**

- We defined electric field  $ec{E}$  at point in space as an electric force per unit charge acting on a test charge
- Now, we define the magnetic flux density  $\vec{B}$  at a point in space in terms of magnetic force that acts on a moving test charge

$$ec{F}_m = qec{u} imes ec{B}$$

- The strength of  $\vec{B}$  is measured in teslas. For a positively charged test particle, the direction of  $\vec{F}$  is that of the cross product containing  $\vec{u}$  and  $\vec{B}$  governed by the right hand rule.
- The strength of  $ec{F}_m$  is given by

$$F_m = q u B \sin heta$$

#### 5.1.1 Magnetic Force on a Current-Carrying Conductor

• For a closed circuit of contour C carrying a current I, the magnetic force is

$$ec{F}_m = I \oint_C dec{l} imes ec{B}$$

#### 5.2 The Biot-Savart Law

• Magnetic flux and magnetic field are related by:

$$\vec{B} = \mu \vec{H}$$

• The Biot-Savart law states that the differential magnetic field  $d\vec{H}$  generated by a steady current I flowing through a differential length  $d\vec{l}$  is:

$$dec{H}=rac{I}{4\pi}rac{dl imes \hat{R}}{R^2}$$

• The magnetic field is orthogonal to the plane containing the direction of the current element and the distance vector

5.2. Magnetic Field Due to Surface and Volume Current Distributions

$$I \ dec{l} = ec{J_S} \ ds = ec{J} \ dV$$

• For an infinitely long wire:

$$ec{B}=rac{\mu_0 I}{2\pi r}\hat{\phi}$$

#### 5.3 Maxwell's Magnetostatic Equations

#### 5.3.1 Gauss' Law for Magnetism

• Just as we had Gauss' Law for Electricity, we can find a magnetic counterpart, the Gauss' Law for Magnetism

$$abla \cdot ec{B} = 0 \Longleftrightarrow \oint_S ec{B} \cdot dec{s} = 0$$

- Magnetic field lines, in contrast to electric field lines, always form continuous closed loops from North to South
- Gauss' Law is constrained to a choice of a Gaussian surface enclosing the charges, similarly, Ampere's Law is constrained to a choice of an Amperian loop encircling the current

#### **5.4 Vector Magnetic Potential**

- We introduce a quantity called the vector magnetic potential A:

 $ec{B} = 
abla imes ec{A}$ 

#### 5.4.1 Vector Poisson's Equation

Given the equations:

$$abla (
abla \cdot A) - 
abla^2 A = \mu J$$

and

 $abla \cdot A = 0$ 

gives us the vector Poisson's equation:

$$abla^2 A = -\mu J$$

Since we can express this equation for each of the coordinate components of A and J, we can write the vector equation:

$$ec{A} = rac{\mu}{4\pi} \int_V rac{ec{J}}{R} \; dV$$

#### 5.4.2 Magnetic Flux

- The magnetic flux  $\Phi$  linking a surface S is defined as the total magnetic flux density passing through it:

$$\Phi = \int_S ec B \cdot dec s$$

• In free space, we modify  $ec{B}=\mu_0ec{H}$  to:

$$ec{B} = \mu_0 ec{H} + \mu_0 ec{M} = \mu_0 (ec{H} + ec{M})$$

#### **5.5 Magnetic Properties of Materials**

• we can classify materials as diamagnetic, paramagnetic, or ferromagnetic

#### 5.5.2 Magnetic Permeability

• In free space,  $ec{B}=\mu_oec{H}$  is modified to:

$$ec{B}=\mu_0ec{H}+\mu_0ec{M}$$

where the magnetization vector  $\vec{M}$  is defined as the vector sum of the magnetic dipole moments of toms contained in a unit volume of the material

- In mose magnetic materials, we have  $ec{M}=\chi_mec{H}$  where  $\chi$  is the magnetic susceptibility of the material

$$\mu=\mu_0(1+\chi_m)$$
 .

#### 5.5.3 Magnetic Hysteresis of Ferromagnetic Materials

- Discusses magnetic domain theory
- Hysteresis means to "lab behind"
  - $\circ$  the existence of a hysteresis loop implies that the magnetization process depends not only on the magnetic field  $\vec{H}$  but also on the magnetic history of the material

#### **5.6 Magnetic Boundary Conditions**

• By analogy of Gauss' Law, we find that

$$\oint_S ec{B} \cdot dec{s} \Longrightarrow B_{1_n} = B_{2_n}$$

• we can further represent that as

$$egin{aligned} \mu_1 H_{1_n} &= \mu_2 H_{1_2} \ & \hat{n} imes (ec{H}_1 - ec{H}_2) = ec{J}_s \end{aligned}$$

- surface currents can exist only on the surfaces of perfect conductors and superconductors. hence, at the interface between media with finite conductivities, we have  $\vec{J_s}=0$  and

$$H_{1t} = H_{2t}$$

• To summarize, we may say that boundary conditions require:

$$ec{B}_{1n} = ec{B}_{2n} ext{ and } rac{ec{B}_{1t}}{\mu_1} = rac{ec{B}_{2t}}{\mu_2}$$

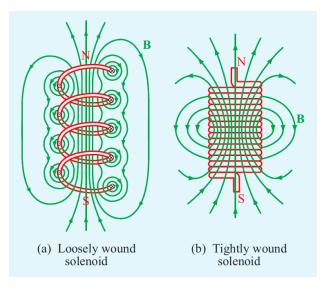
#### 5.7 Inductance

٠

• a typical inductor consists of multiple turns of wire helically coiled around a cylindrical core, such a structure is called a solenoid

#### 5.7.1 Magnetic Field in a Solenoid

$$ec{H} = \hat{z} rac{Ia^2}{2(a^2+z^2)^{3/2}}$$



$$ec{B}=rac{\mu NI}{l}\hat{z}$$

- self-inductance is the magnetic flux linkage of a coil or circuit with itself
- mutual inductance involves the magnetic flux linkage in a circuit due to the magnetic field generated by a current in another one

#### 5.7.2 Self-Inductance of a Solenoid

$$\Phi = \int_S ec{B} \cdot dec{s}$$

- Magnetic flux linkage  $\Lambda$  is defined as the total magnetic flux linking a given circuit or conducting structure

$$\Lambda = N \Phi = \mu rac{N^2}{l} IS$$

- The self-inductance of any conducting structure is defined as the ratio of the magnetic flux linkage  $\Lambda$  to the current I flowing through the structure

$$L = rac{\Lambda}{I}$$

#### 5.7.3 Self-Inductance of Other Conductors

• for a two-conductor configuration for either two parallel wires or a coaxial wire, the inductance is given by:

$$L=rac{\Lambda}{I}=rac{\Phi}{I}=rac{1}{I}\int_{S}ec{B}\cdot dec{s}$$

#### 5.7.4 Mutual Inductance

$$L_{12} = rac{\Lambda_{12}}{I_1} = rac{N_{12}}{I_1} \int_{S_2} ec{B}_1 \cdot dec{s}$$

### 5.8 Magnetic Energy

$$W_m=rac{1}{2}Li^2$$

- this is the magnetic energy stored in the inductor
- we can also define the magnetic energy density

$$w_m=rac{W_m}{V}=rac{1}{2}\mu H^2$$
 .

- for any volume V containing a material with permeability  $\mu$  the total magnetic energy stored in a magnetic field is

$$W_m = {1\over 2} \int_V \mu H^2 \; dV$$



## Maxwell's Equations for Time-Varying Fields

#### 6.1 Faraday's Law

- Faraday hypothesized that if a current produces a magnetic field, then the converse should also be true:
  - A magnetic field should produce a current in a wire

$$\Phi = \int_S ec{B} \cdot dec{s}$$

- current is only induced when the magnetic flux changes, and the direction of the current is dependent on whether the flux is increasing or decreasing
- when a galvanometer detects the flow of current through the coil, it means that a voltage has been induced across the galvanometer terminals
  - $\circ$  called the electromotive force  $V_{emf}$  (it's a voltage not a force)

$$V_{emf} = -Nrac{d\Phi}{dt} = -Nrac{d}{dt}\int_{S}ec{B}\cdot dec{s}$$

- an emf can be generated in a closed conducting loop under any of the following three conditions:
  - a time-varying magnetic field linking a stationary loop, called a transformer emf
  - a moving loop with a time-varying area in a static field, called the motional emf
  - a moving loop in a time-varying field
- total emf is given by:

$$V_{emf} = V_{emf}^{tr} + V_{emf}^m$$

# 6.2 Stationary Loop in a Time-Varying Magnetic Field

$$V^{tr}_{emf} = -N \int_S rac{\partial ec{B}}{\partial t} \cdot dec{s}$$

• the transformer emf is the voltage difference that would appear across the small opening in terminals

$$abla imes ec{E} = -rac{\partial ec{B}}{\partial t}$$

### 6.3 The Ideal Transformer

$$rac{V_1}{V_2}=rac{N_1}{N_2} ext{ and } rac{I_1}{I_2}=rac{N_2}{N_1}$$

# 6.4 Moving Conductor in a Static Magnetic Field

• the field generated by the motion of the charged particle is called a motional electric field

$$ec{E}_m = rac{ec{F}_m}{q} = ec{u} imes ec{B}$$

• in general, if any segment of a closed circuit with contour C moves with a velocity  $\vec{u}$  across a static magnetic field  $\vec{B}$  then the induced motional emf is given by:

$$V^m_{emf}=\oint_C (ec{u} imesec{B})\cdot dec{l}$$

### 6.5 The Electromagnetic Generator

• the magnetic field is

$$ec{B}=\hat{z}B_{0}$$

- as the loop rotates with an angular velocity  $\omega$  about its own axis, the segments move with:

$$ec{u}=\hat{n}\omegarac{w}{2}$$

$$\hat{n} imes \hat{z} = \hat{x} \sin lpha$$

• we then obtain the result

$$V^m_{emf} = w l \omega B_0 \sin lpha = A \omega B_0 \sin lpha$$

- the angle  $\alpha$  is realted to  $\omega$  by

$$lpha=\omega t+C_0$$

## 6.6 Moving Conductor in a Time-Varying Magnetic Field

$$V_{emf} = -rac{d\Phi}{dt} = -rac{d}{dt}\int_{S}ec{B}\cdot dec{s}$$

### 6.7 Displacement Current

• Ampere's Law in differential form is given by

$$abla imes ec{H} = ec{J} + rac{\partial ec{D}}{\partial t}$$

- we integrate both sides over an arbitrary open surface S and contour C.
  - $\circ$  the surface integral of  $ec{J}$  is the conduction current  $I_c$  flowing through S, and
  - $\circ$  the surface integral of  $abla imes ec{H}$  becomes a line interal of  $ec{H}$  over C

$$\oint_C ec{H} \cdot dec{l} = I_c + \int_S rac{\partialec{D}}{\partial t} \cdot dec{l}$$

- the second term on the right has units of amperes obviously, and is proportional to the time derivative of the electric flux density  $\vec{D}$ 
  - $\circ$  this is called the displacement current  $I_d$

$$I_d = \int_S = ec{J_d} \cdot dec{s} = \int_S rac{\partialec{D}}{\partial t} \cdot dec{s}$$

• from the above two equations, we can say that

$$\oint_C ec{H} \cdot dec{l} = I_c + I_d = I$$

#### 6.9 Charge-Current Continuity Relation

- we define I as the net current flowing across S out of V
  - $\circ$  thus, I is the negative rate of change of Q

$$I=-rac{dQ}{dt}=-ddt\int_V
ho_V\,dV$$

- the current I is also defined as the outward flux of the current density  $\vec{J}$  through the surface S

$$\oint_S ec{J} \cdot dec{s} = -rac{d}{dt} \int_V 
ho_V \, dV$$

- apply the divergence theorem and convert the surface integral of  $\vec{J}$  into a volume integral of its divergence

$$\oint_S ec{J} \cdot dec{s} = \int_V 
abla \cdot ec{J} \ dV = -rac{d}{dt} \int_V 
ho_V \ dV$$

- we can move the time derivative inside the integral and express as a partial derivative of  $\rho_V$ , then drop both volume integrals

$$abla \cdot ec J = -rac{\partial 
ho_V}{\partial t}$$
 .

- this is known as the charge-current continuity relation, or the **charge continuity equation**.
- another expression for Kirchhoff's current law:

$$\oint_S ec{J} \cdot dec{s} = 0$$